

# Optimization of beam coupling impedance of button and stripline BPMs

Shalva Bilanishvili

December 12, 2024



The project has received funding from the European Union's Horizon 2020 research and innovation programme under grant agreement [871072]

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#### **Outline**

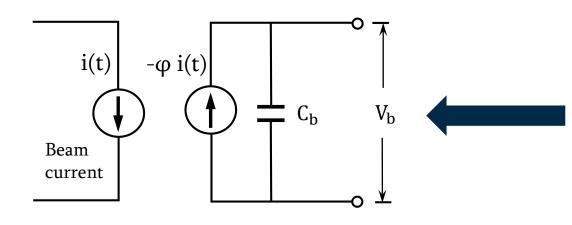


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- Introduction
- Motivation
- Button BPMs
- Stripline BPMs
- Simulation and Analysis
- Conclusions

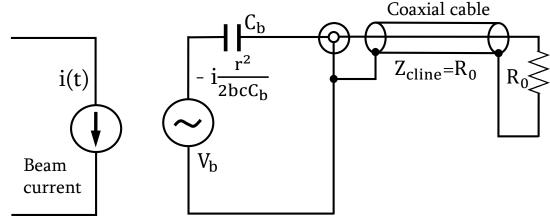
#### **Beam Position Monitors**





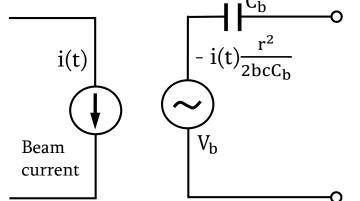
The usual equivalent circuit representation of the ordinary button pick-up is a current generator of the same value of the current intercepted fraction  $\Delta i_b$ , shunted by the electrode capacitance to ground  $C_{\rm b}$ 

Now, let us rearrange the above equivalent scheme into a scheme with the voltage generator



1(L) 2bcC<sub>b</sub> https://doi.org/10.1016/S0168-9002(97)01083-8

The button is now terminated into the resistor  $R_0$ , via a coaxial cable with the characteristic impedance  $Z_0=R_0$  in this corresponding equivalent scheme



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#### **Beam Transfer Impedance**



In the frequency domain, the voltage at the termination load  $R_0$  is given by:

$$V_0(\omega) = V_b(\omega) \frac{R_0}{R_0 + 1/j\omega C_b} = I(\omega) \frac{r^2}{2bc} \frac{1}{C_b} \frac{j\omega R_0 C_b}{1 + j\omega R_0 C_b}$$

Here  $V_0(\omega)$  and  $I(\omega)$  are the spectral densities of the load voltage  $V_0(t)$  and beam current i(t), respectively. Then, according to the definition, the transfer (signal) impedance is given by:

$$Z_b(\omega) \equiv \frac{V(\omega)}{I(\omega)} = R_0 \frac{j\omega r^2/2bc}{1 + j\omega R_0 C_b}$$

With definitions  $\omega_1 = 1/R_0C_b$  and  $\omega_2 = c/2r$ , we can write

$$Z_b(\omega) = \varphi R_0 \left(\frac{\omega_1}{\omega_2}\right) \frac{j\omega/\omega_1}{1 + (\omega/\omega_1)^2}$$

$$\operatorname{Re} Z_b(\omega) = \varphi R_0 \left(\frac{\omega_1}{\omega_2}\right) \frac{(\omega/\omega_1)^2}{1 + (\omega/\omega_1)^2}$$

$$Im Z_b(\omega) = \varphi R_0 \left(\frac{\omega_1}{\omega_2}\right) \frac{(\omega/\omega_1)}{1 + (\omega/\omega_1)^2}$$

#### **Beam Coupling Impedance**



In order to get the coupling impedance, we calculate the average power dissipated at the external termination  $R_0$ 

$$\langle P(\omega) \rangle = \frac{1}{2} |V_0|^2 \frac{1}{R_0} = \frac{1}{2} |I|^2 R_0 \varphi^2 \left(\frac{\omega_1}{\omega_2}\right)^2 \frac{(\omega/\omega_1)^2}{1 + (\omega/\omega_1)^2}$$

This power is provided by the beam and can also be written as:

$$\langle P(\omega) \rangle = \frac{1}{2} |I|^2 Re\{Z_l(\omega)\}$$

By comparing these two equations we can get the expression for the real part of the longitudinal coupling impedance:

$$\mathbf{Re} Z_l(\omega) = R_0 \varphi^2 \left(\frac{\omega_1}{\omega_2}\right)^2 \frac{(\omega/\omega_1)^2}{1 + (\omega/\omega_1)^2}$$

In a causal system the real and imaginary  $\operatorname{Re} Z_l(\omega) = R_0 \varphi^2 \left(\frac{\omega_1}{\omega_2}\right)^2 \frac{(\omega/\omega_1)^2}{1 + (\omega/\omega_1)^2}$  parts of the Fourier transform of the pulse response function are related to each other by a Hilbert transforms pair

$$\mathbf{Im} Z_l(\omega) = R_0 \varphi^2 \left(\frac{\omega_1}{\omega_2}\right)^2 \frac{\omega/\omega_1}{1 + (\omega/\omega_1)^2}$$

The annular cut also contributes to the coupling impedance, which can be estimated analytically at low frequencies.

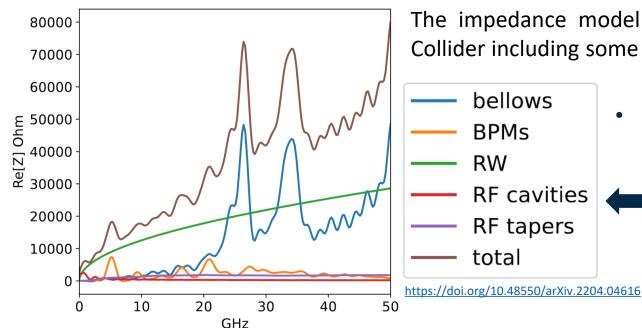
$$Z_{l cut}(\omega) = j \frac{Z_0 \omega (r + \omega)^3}{8cb^2 \left( \ln \left[ \frac{32(r + \omega)}{\omega} \right] - 2 \right)}$$

https://doi.org/10.1109/PAC.1995.505888

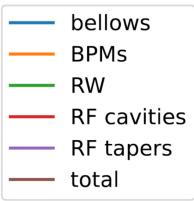
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#### **Some Examples**





The impedance model evaluated for Future Circular  $e^+e^-$ Collider including some of the vacuum chamber components.



It is important to note that, the beam position monitors produce second largest contribution in the total impedance budget up to 10 GHz).

> In the PEP-II B-Factory, some of the upper button electrodes heated up enough to fall off their mounts. The upper electrode fell onto the lower electrode.



It shorted the underlying electrode and also became a large obstacle for the beam fields, increasing the current though the lower electrode. This then melted the feed-through causing a vacuum breach.

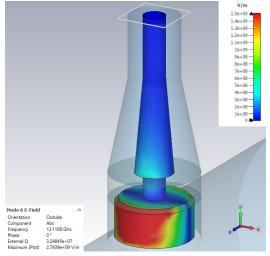
The origin of the heating was the wake field generated by an intense short bunch passing by the vacuum chamber discontinuity due to a BPM button.



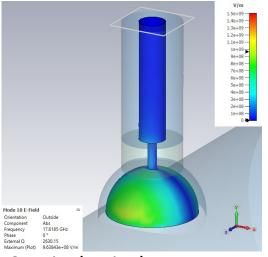
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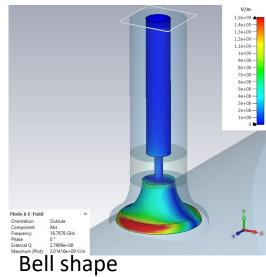
# **Button BPM Electrode Design Variations**





3.2e+09 — 3e+09 — 2.8e+09 — 2.4e+09 — 2.2e+09 — 2.e+09 — 1.6e+09 — 1.4e+09 — 1.2e+09 — 1.e+09 — 8e+08 — 6e+08 — 4e+08 — 2e+08 — 6e+08 Component Frequency Phase External Q Abs 15.8344 GHz

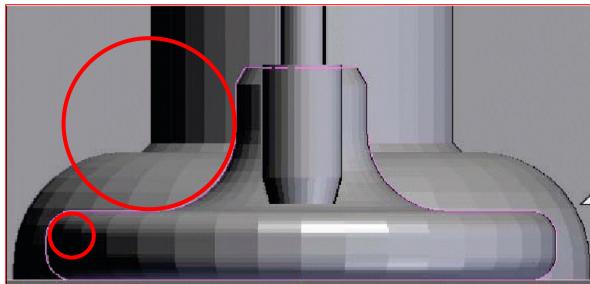


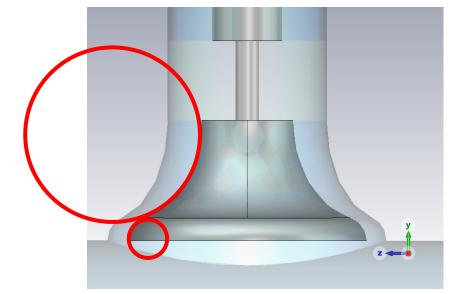


Cylindrical

Conical

Semispherical

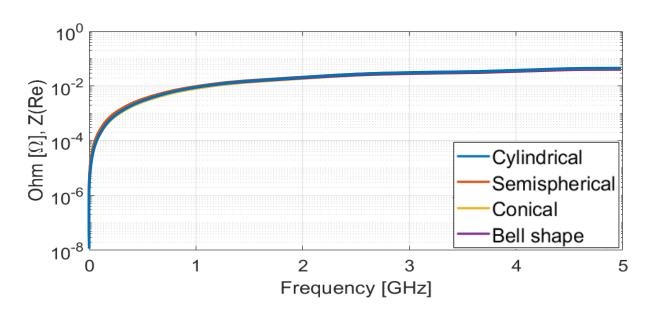


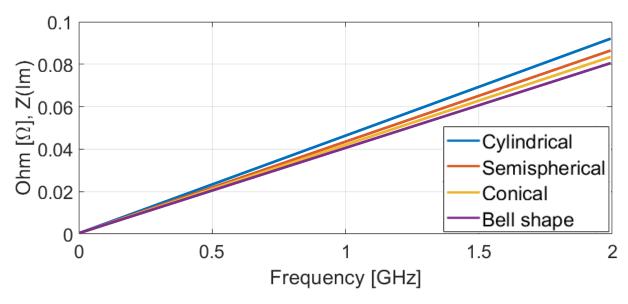


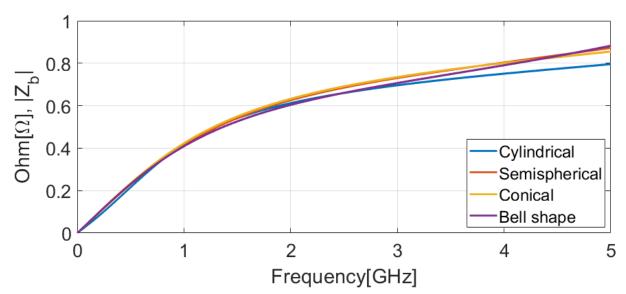
https://accelconf.web.cern.ch/ICAP2009/papers/thpsc057.pdf

### **Button BPM Electrode Design Variations**





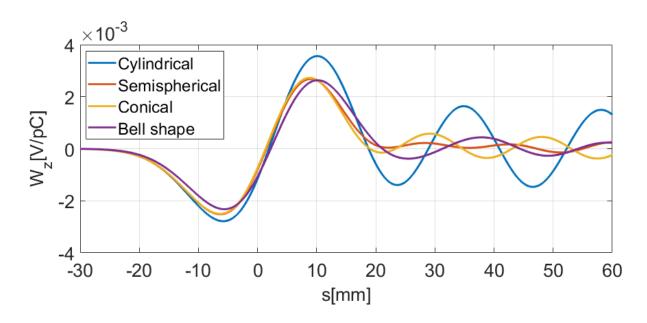


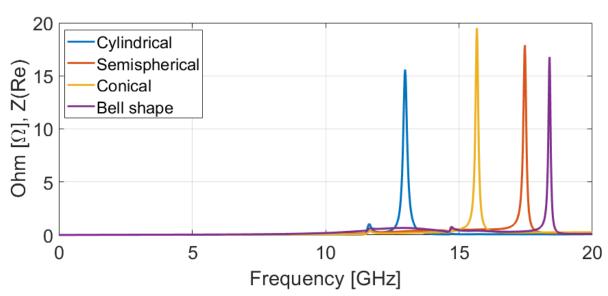


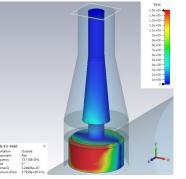
• It was examined how the button geometry can be modified to minimize the harmful effect of the resonance peak narrow-band impedance, while keeping the transfer impedance, button capacitance, and gap between the electrode and vacuum chamber on the same level.

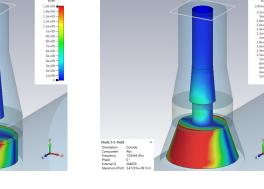
https://doi.org/10.1088/1742-6596/2687/7/072004

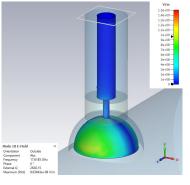
#### **Design Optimization Results for Button BPMs**

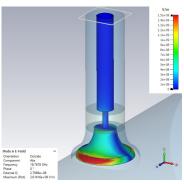








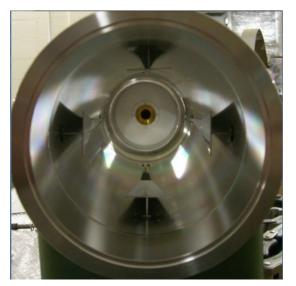


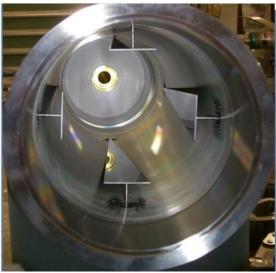


- A choice of button geometry can significantly affect the narrow-band beam coupling longitudinal impedance of BPMs in vacuum chambers.
- The issues of heating effects and coupled-bunch instabilities, caused by BPMs' contribution to both broad-band and narrow-band impedances can be effectively addressed.

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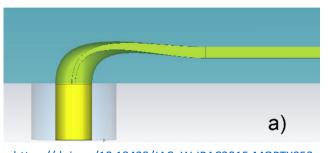
### **Stripline BPM Design Overview**



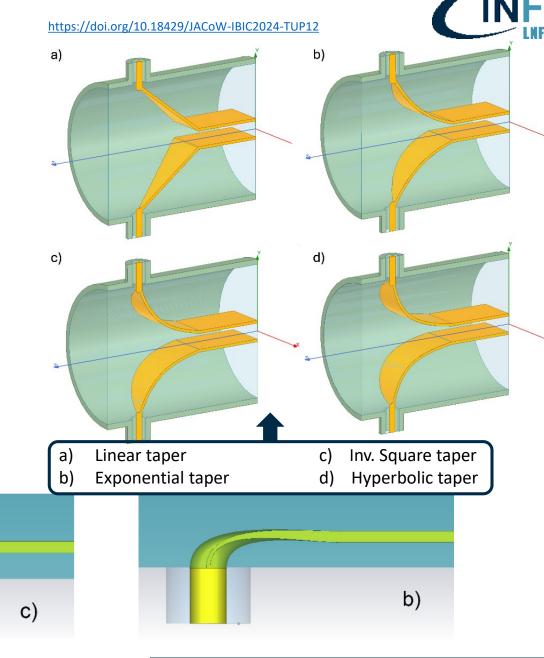


https://doi.org/10.1103/PhysRevAccelBeams.19.021003

There were several studies dedicated to stripline BPM's to use nonlinear tapered transition for stripline to feed-through to optimize the characteristic impedance



https://doi.org/10.18429/JACoW-IPAC2015-MOPTY053



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#### **Stripline BPM Design Overview**

INFN

http://cds.cern.ch/record/210347

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CM-P00061004

#### Impedance of Slowly Tapered Structures

Kaoru Yokoya\*



Function that we used for exponential tapering

https://doi.org/10.1103/PhysRevSTAB.10.074402

#### PHYSICAL REVIEW SPECIAL TOPICS - ACCELERATORS AND BEAMS 10, 074402 (2007)

#### Transverse impedance of tapered transitions with elliptical cross section

B. Podobedov and S. Krinsky

Brookhaven National Laboratory, Upton, New York 11973, USA
(Received 23 May 2007; published 30 July 2007)

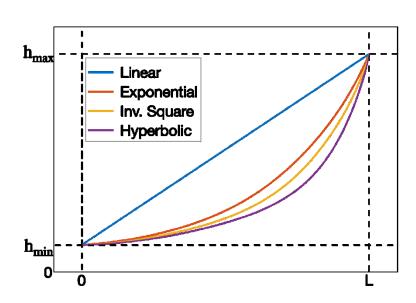


Function that was used for inv. square tapering

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#### **Design Optimization Studies for Stripline BPMs**





$$h(z) = h_{min}(1 + (h_{max}/h_{min} - 1)z/L)$$

$$h(z) = h_{min} (h_{max}/h_{min})^{z/L}$$

$$h(z) = \frac{h_{min}}{(1 + ((h_{min}/h_{max})^{1/2} - 1)z/L)^2}$$

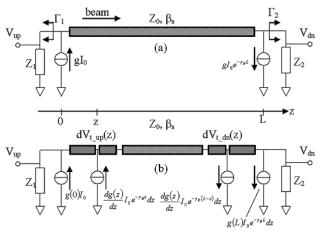
$$h(z) = \frac{h_{min}}{1 + ((h_{min}/h_{max}) - 1)z/L},$$

Linear

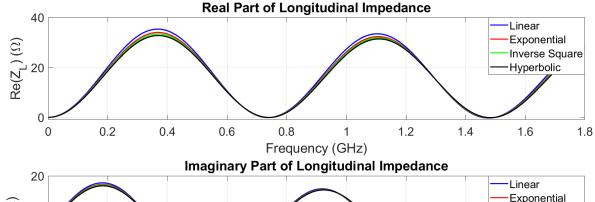
**Exponential** 

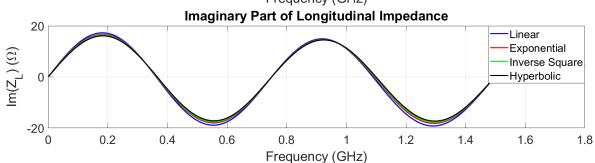
Inv. Square

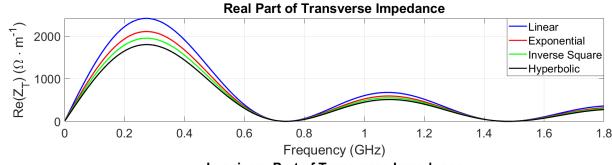
Hyperbolic

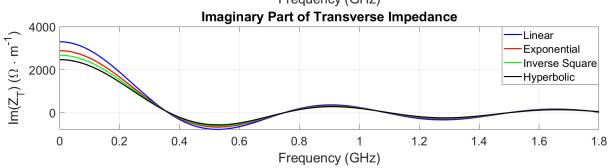


https://doi.org/10.1103/PhysRevSTAB.13.111002





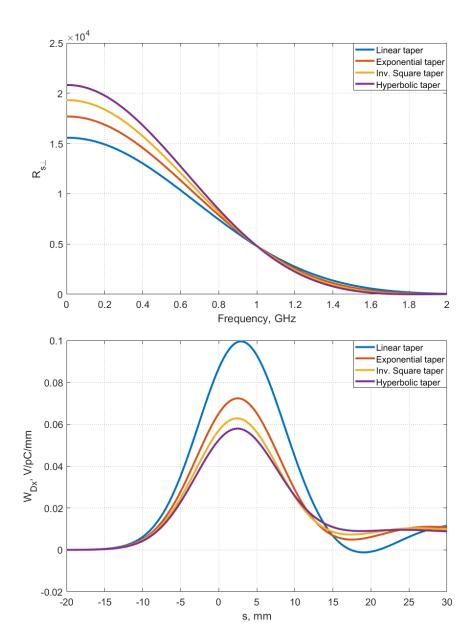


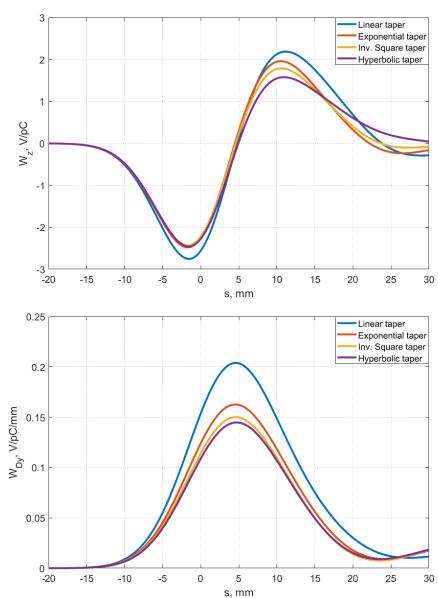


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## **Design Optimization Results for Stripline BPMs**







#### **Design Optimization Results for Stripline BPMs**



	Wake-Loss-Factor	Unit
Linear:	1.380018e+00	V/pC
Exponential:	1.196212e+00	V/pC
Inverse Square:	1.203277e+00	V/pC
Hyperbolic:	1.209588e+00	V/pC

	Kick-Factor x	Kick-Factor y	Unit
Linear:	-6.840428e-02	-1.327580e-01	V/pC/mm
Exponential:	-5.037268e-02	-1.068364e-01	V/pC/mm
<b>Inverse Square:</b>	-4.401820e-02	-9.837109e-02	V/pC/mm
Inverse Square: Hyperbolic:	-4.080152e-02	-9.439061e-02	V/pC/mm

- For the stripline devices (kickers, BPMs) with the same total strip length and gap between the strips and having the same characteristic impedance, the nonlinear tapering can add several beneficial features.
- In particular, it can enhance the stripline functionality, by increasing the transverse shunt impedance.
- In addition, the nonlinear tapering results in a significant reduction of the beam coupling impedance and respective loss and kick factors.
- Overall, the choice of tapering function and structure aspect ratio significantly impacts impedance characteristics and device efficiency.

#### **Conclusions**



- Adjusting or choosing BPM button geometry can reduce harmful effects and shift narrow band resonance peak impedances towards higher frequencies, while keeping the transfer impedance, button capacitance, and gap between the electrode and vacuum chamber on the same level.
- For striplines (such as kickers and BPMs), using nonlinear tapers can improve performance by increasing transverse shunt impedance and also reduce beam coupling impedance, with the same total strip length and gap between the strips and having the same characteristic impedance.
- Overall, both button electrode geometry and stripline tapering choices can play an important role in improving efficiency and stability of these devices.



# Thank you